

# Natural Entrainment of Collocated Mechanical Systems via Decentralized Multi-agent Feedback

Y. Futakata

Information Physics & Computing  
The University of Tokyo  
futakata\_yoshiaki@ipc.i.u-tokyo.ac.jp

T. Iwasaki

Mechanical & Aerospace Engineering  
University of California, Los Angeles  
tiwasaki@ucla.edu

**Abstract**—This paper considers the design of feedback controllers for a class of collocated mechanical systems, aiming to achieve a natural mode of oscillation as a stable limit cycle of the closed-loop system. Motivated by recent results on central pattern generators, the controller is formed as a collection of multiple identical agents without direct communications to each other. Each agent is described as a transfer function followed by a static nonlinearity, and is placed between a sensor/actuator pair. Several simple linear dynamics are considered for the agents. The method of multivariable harmonic balance (MHB) is used to obtain approximate conditions for achieving entrainment to a natural oscillation, and numerical examples suggest that the proposed design conditions are fairly reliable in spite of sinusoidal approximations adopted in the MHB method. It is found that a band-pass filtering property is essential for the linear dynamics of each agent to achieve entrainment to an arbitrarily specified mode of natural oscillation.

## I. INTRODUCTION

Resonance, or natural motion in general, can often be exploited for increased efficiency when controlling a mechanical system to achieve oscillatory behaviors. Rhythmic body movements during animal locomotion are known to be controlled by central pattern generators (CPGs), a group of neurons that are interconnected in a specific manner [1], [2]. Recent studies on biological control systems have shown that feedback controllers with architectures borrowed from CPGs can achieve entrainment to the resonance of a single degree-of-freedom (DOF) mechanical system [3]–[7], and more generally, entrainment to a natural mode of oscillation for multi-DOF systems [8], [9]. While the dynamics of a CPG, which is a nonlinear oscillator by itself, can be fairly complex, it is suggested [3], [7] that the mechanisms underlying the natural entrainment are rather simple — locally negative damping supplied through positive rate feedback or negative integral feedback, with saturation.

To explain the idea, consider a single-DOF system

$$m\ddot{x} + d\dot{x} + kx = u$$

where  $m$ ,  $k$ , and  $d$  are the mass, stiffness, and damping,  $x(t)$  is the displacement, and  $u(t)$  is the force input. The objective is to design a feedback controller to achieve an oscillation at the natural frequency  $\sqrt{k/m}$ . The simplest solution would be  $u(t) = d\dot{x}$  to cancel the damping, but this

This work is supported in part by NSF No.0654070, by ONR MURI Grant N00014-08-1-0642, and by The Ministry of Education, Science, Sport, and Culture, Japan under Grant 21246067.

does not result in structurally stable oscillation. The idea of positive rate feedback with saturation is to use the control law  $u = \psi(\dot{x})$  where  $\psi$  is a saturation nonlinearity with the slope at the origin greater than  $d$ . Such feedback will provide locally negative damping to destabilize the origin. On the other hand, the saturation effect will reduce the negative damping in the region away from the origin, making every trajectories bounded. As a result, the trajectory converges to a limit cycle on which the negative damping supplied by the control input just cancels the mechanical damping, leading to an oscillation at (or near) the natural frequency. The idea of negative integral feedback with saturation is less obvious in the time domain, but it is easy to see that  $u = -\psi(\int x dt)$  would have an effect similar to  $u = \psi(\dot{x})$  in the frequency domain, at least for the case where  $\psi$  is a linear function.

Motivated by the natural entrainment results, this paper explores the potential of multi-agent controllers with simple nonlinear dynamics to achieve entrainment to a prescribed mode of natural oscillation. In particular, we consider the class of mechanical systems arising from body biomechanics, where stiffness, damping, sensors, and actuators are all collocated. An “agent” is described by  $u_i = g_i\psi(r(s)h_i y_i)$  and is placed between each mechanical (input,output) pair  $(u_i, y_i)$  as a feedback controller, where  $g_i$  and  $h_i$  are the actuator and sensor gains,  $r(s)$  is a single-input-single-output (SISO) transfer function, and  $\psi$  is a saturation-type nonlinearity. We consider the differentiator and integrator for  $r(s)$ , as well as their extensions to the low-pass, high-pass, and band-pass filters. We shall investigate whether a set of agents, without direct communications to each other, is able to achieve natural entrainment, in line with a previous research [8] that revealed natural entrainment capability of decentralized CPGs. A systematic method will be developed for designing decentralized multi-agent controllers so that the multivariable harmonic balance (MHB) equation [10], [11] admits a solution at or near one of the natural modes of oscillation. Numerical examples will then validate the accuracy and reliability of the proposed design method. All proofs except the one for Theorem 1 are omitted due to the page limitation.

We use the following notation. Let  $j := \sqrt{-1}$ . The symbols  $\mathbb{R}_+$ , and  $\mathbb{I}_n$  denote the set of real positive numbers and the set of integers from 1 to  $n$ , respectively. For a vector  $v \in \mathbb{R}^n$  with the  $i^{\text{th}}$  entry  $v_i$ , the  $n \times n$  diagonal matrix whose  $(i, i)$  entry is  $v_i$ , is denoted by  $\text{diag}(v)$  or  $\text{diag}(v_1, \dots, v_n)$ .

For sets  $X$  and  $Y$ , the set  $X/Y$  is defined as a set of elements which belong to  $X$  but not to  $Y$ .

## II. PROBLEM FORMULATION AND APPROACH

### A. Problem Statement

Consider the class of mechanical systems given by

$$J\ddot{x} + D\dot{x} + Kx = Bu, \quad y = Cx, \quad (1)$$

where  $x(t) \in \mathbb{R}^n$ , and  $u(t), y(t) \in \mathbb{R}^m$  are the generalized coordinates, inputs, and outputs, respectively, and  $J$ ,  $D$ , and  $K$  are the inertia, damping, and stiffness matrices, respectively. Recall that a natural mode of the system (1) is defined by a pair of the natural frequency  $\omega_\ell \in \mathbb{R}_+$  and mode shape  $\xi_\ell \in \mathbb{R}^n$  satisfying  $(K - \omega_\ell^2 J)\xi_\ell = 0$  with  $\xi_\ell \neq 0$ . There are  $n$  natural modes,  $(\omega_\ell, \xi_\ell)$  with  $\ell \in \mathbb{I}_n$ , for the  $n$ -DOF system. We define the  $\ell^{\text{th}}$  mode shape of the output  $\hat{y}_\ell := C\xi_\ell$  and denote its  $i^{\text{th}}$  entry by  $\hat{y}_{\ell i}$ . The following assumption is enforced throughout the paper.

*Assumption 1:* The matrices  $B$  and  $C$  are square ( $m = n$ ) nonsingular, and satisfy  $C = B^\top$ . The matrices  $J$ ,  $D$ , and  $K$  are symmetric positive definite. The stiffness matrix is given by

$$K = BKC, \quad \mathcal{K} := \text{diag}(k_1, \dots, k_n).$$

for some scalars  $k_i \in \mathbb{R}_+$  with  $i \in \mathbb{I}_n$ . The damping matrix is given by  $D = \rho K$  for some  $\rho \in \mathbb{R}_+$ . The natural frequencies are distinct and are arranged in the ascending order:  $0 < \omega_1 < \omega_2 < \dots < \omega_n$ . During each natural mode of oscillation, no output variable remains stationary:  $\hat{y}_{\ell i} \neq 0$  for all  $\ell, i \in \mathbb{I}_n$ .

The conditions in Assumption 1 are motivated by biological systems where the actuators, sensors, and stiffness and damping elements are all located at the same position. In addition, Rayleigh damping  $D = \rho K$  and distinct natural frequencies are assumed so that the analysis becomes simple enough to provide insights into the natural entrainment mechanism. We shall refer to the system (1) satisfying Assumption 1 as the *collocated* system.

We consider a set of  $n$  identical agents to control the system. Each agent is an SISO system described as a linear transfer function  $r(s)$  followed by a static nonlinearity  $\psi$ , and is placed between each sensor/actuator pair as a feedback controller:

$$u_i = g_i \psi(q_i), \quad q_i = r(s) h_i y_i$$

where  $g_i$  and  $h_i$  are feedback gains, and  $q_i$  is an internal variable. The static nonlinearity  $\psi$  is a sigmoid function having the following properties:

- $\psi$  is odd, bounded, and strictly increasing.
- $\psi(x)$  is strictly concave on  $x > 0$ , and  $\psi'(0) = 1$ .

We denote by  $\Psi$  the class of functions satisfying these conditions. For example,  $\tanh(x)$  belongs to this class. The set of  $n$  agents can be written as

$$u = G\Psi(q), \quad q = r(s)Hy, \quad (2)$$

where  $g, h \in \mathbb{R}^n$  and  $q(t) \in \mathbb{R}^n$  are the vectors defined by stacking  $g_i, h_i$ , and  $q_i(t)$  in columns, respectively, and

$$G := \text{diag}(g), \quad H := \text{diag}(h), \quad \Psi := \psi I.$$

To seek an essential mechanism for natural entrainment, we consider five different types of the transfer function  $r(s)$ :

D	I	LPF	HPF	BPF
$s$	$\frac{1}{s}$	$\frac{\omega_o}{s + \omega_o}$	$\frac{s}{s + \omega_o}$	$\frac{2\omega_o s}{(s + \omega_o)^2}$

where D, I, LPF, HPF and BPF stand for the differentiator, integrator, low-pass filter, high-pass filter and band-pass filter, respectively. Throughout the paper, we restrict our attention to the case where  $\omega_o$  is positive,  $g_i$  and  $h_i$  are nonzero for all  $i \in \mathbb{I}_n$ , and  $\psi \in \Psi$ . In addition, we consider positive  $h_i$  only, as its sign can be absorbed into  $g_i$ . Let us denote by  $\mathcal{R}$  the collection of the five transfer functions.

With the controller (2) and the mechanical system (1), the closed-loop system is given by  $\mathcal{G}(s, \Psi)z = 0$  where

$$\mathcal{G}(s, \Psi) := \begin{bmatrix} Js^2 + Ds + K & -BG\Psi \\ r(s)HC & -I \end{bmatrix}, \quad z := \begin{bmatrix} x \\ q \end{bmatrix}. \quad (3)$$

We would like to solve the following problems posed on (3):

- 1) Characterize design conditions on the controller such that the closed-loop system has a prescribed mode of natural oscillation as a stable limit cycle.
- 2) Compare the design conditions for the five transfer functions  $r(s) \in \mathcal{R}$ , and determine what agent dynamics is essential for achieving natural entrainment.

These problems will be addressed in an approximate setting as described in the next section.

### B. Multivariable Harmonic Balance

Assume that the closed-loop system (3) has a periodic solution  $z(t)$  with a frequency  $\omega$ . We shall approximate  $z(t)$  by a sinusoidal signal  $z(t) = \Re[\hat{z}e^{j\omega t}]$ , where  $\hat{z} \in \mathbb{C}^{2n}$  is the phasor of  $z(t)$ . The nonlinearity  $\psi$  may be approximated by the describing function when its input signal is a sinusoid:

$$\begin{aligned} \psi(v) &\cong \kappa(\alpha)v \quad \text{for } v := \alpha \sin \omega t, \\ \kappa(\alpha) &:= \frac{4}{\alpha\pi} \int_0^{\pi/2} \psi(\alpha \sin \theta) \sin \theta d\theta. \end{aligned} \quad (4)$$

It was shown in [7] that  $\kappa(\alpha)$  for  $\psi \in \Psi$  takes a value in the interval  $(0, 1)$  when  $\alpha > 0$ , and  $\kappa$  is strictly decreasing from 1 to 0 on  $\alpha > 0$ .

With the sinusoidal approximations and describing functions, the equation for the closed-loop system (3) reduces to the multivariable harmonic balance (MHB) equation:

$$\mathcal{G}(j\omega, \Phi)\hat{z} = 0, \quad (5)$$

where  $\Phi$  is an amplitude-dependent matrix gain that approximates the nonlinear function  $\Psi$  for sinusoidal inputs:

$$\Phi := \text{diag}(\kappa_1, \dots, \kappa_n), \quad \kappa_i := \kappa(|\hat{q}_i|),$$

where  $\hat{q}_i$  is the phasor of  $q_i(t)$ . For a set of fixed amplitudes  $|\hat{q}_i|$ , let us define the associated linear system by  $\mathcal{G}(s, \Phi)z = 0$ , or equivalently,

$$\begin{aligned} A(s)x &= 0, \\ A(s) &:= Js^2 + Ds + K - r(s)BG\Phi HC. \end{aligned} \quad (6)$$

The basic idea of the MHB method [10] is the following.

- If  $(\omega, \hat{z})$  satisfies (5), then the system (3) is expected to possess oscillatory trajectory  $z(t) \cong \Re[\hat{z}e^{j\omega t}]$ .
- The estimated oscillation is expected to be stable if the associated linear system (6) is marginally stable with  $\pm j\omega$  being the only eigenvalues on the imaginary axis.

We call  $(\omega, \hat{z})$  a *stable solution* of (5) if it satisfies (5), and (6) satisfies the marginal stability condition. The controller (2) is said to *assign*  $(\omega, \hat{x})$  a (stable) oscillation profile if  $(\omega, \hat{x}, \hat{q})$  is a (stable) solution of (5) for some  $\hat{q}$ .

### III. NATURAL ENTRAINMENT ANALYSIS

#### A. The MHB condition

Consider the mechanical system (1) driven by the feedback controller (2). We would like to characterize the set of controller parameters for which the MHB equation (5) has a stable solution  $(\omega, \hat{x}) = (\omega_\ell, \xi_\ell)$  at a prescribed mode  $\ell \in \mathbb{I}_n$ . However, it turns out that such set is empty except for a special case (I with  $\ell = 1$ ) and therefore we consider a relaxation of the natural entrainment condition. In particular, we enforce the mode shape but not the frequency, and seek a stable solution  $(\omega, \hat{x})$  such that  $\hat{x} = \xi_\ell$  and  $\omega \cong \omega_\ell$ .

First note that the phasor equation for the mechanical system is given by

$$(K + j\omega D - \omega^2 J)\hat{x} = B\hat{u}, \quad \hat{y} = C\hat{x}.$$

When  $\hat{x} = \xi_\ell$ , this reduces to

$$\left[ \left( 1 - \left( \frac{\omega}{\omega_\ell} \right)^2 \right) K + j\omega D \right] \xi_\ell = B\hat{u}, \quad \hat{y} = C\xi_\ell.$$

Exploiting the collocated structure,

$$\left[ 1 - \left( \frac{\omega}{\omega_\ell} \right)^2 + j\omega\rho \right] k_i \hat{y}_{\ell i} = \hat{u}_i.$$

Thus, the mechanical system imposes the following input-output relationship when the mechanical variables oscillate sinusoidally with a mode shape  $\xi_\ell$ :

$$\frac{\hat{y}_{\ell i}}{\hat{u}_i} = \frac{p_\ell(\omega)}{k_i}, \quad p_\ell(\omega) := \frac{\omega_\ell}{2\omega(j\zeta_\ell - \varpi_\ell)}, \quad (7)$$

$$\zeta_\ell := \frac{\rho\omega_\ell}{2}, \quad \varpi_\ell := \frac{1}{2} \left( \frac{\omega}{\omega_\ell} - \frac{\omega_\ell}{\omega} \right). \quad (8)$$

For later reference, we define  $\zeta_\ell$  and  $\varpi_\ell$  for  $\ell \in \mathbb{I}_n$  and  $\ell = o$ . The phasor equation for the controller, after quasi-linearization via describing functions, is given by

$$\hat{u} = G\Phi\hat{q}, \quad r(j\omega)HC\hat{x} = \hat{q},$$

or equivalently,

$$\frac{\hat{u}_i}{\hat{y}_i} = g_i h_i \kappa(|\hat{q}_i|) r(j\omega), \quad \hat{q}_i = r(j\omega) h_i \hat{y}_i. \quad (9)$$

From (7) and (9), the MHB equation is given by

$$g_i h_i \kappa(|\hat{q}_i|) p_\ell(\omega) r(j\omega) / k_i = 1.$$

We denote the real part and imaginary parts of  $1/r(j\omega)$  and  $1/(p_\ell(\omega)r(j\omega))$  by

$$\sigma + j\lambda := \frac{1}{r(j\omega)}, \quad \phi_\ell + j\theta_\ell := \frac{1}{p_\ell(\omega)r(j\omega)}. \quad (10)$$

For each case of  $r(s)$ , the parameters  $\sigma$  and  $\lambda$  are given by

	D	I	LPF	HPF	BPF
$\sigma$	0	0	1	1	1
$\lambda$	$-\frac{1}{\omega}$	$\omega$	$\frac{\omega}{\omega_o}$	$-\frac{\omega_o}{\omega}$	$\varpi_o$

(11)

The following result summarizes the above development and provides a necessary and sufficient condition for satisfaction of the MHB equation in a form suitable for analysis and synthesis.

*Lemma 1:* Consider the mechanical system (1) and controller (2), where  $r(s)$  is an arbitrary transfer function that possibly depends on a parameter  $\omega_o$ . Let  $\ell \in \mathbb{I}_n$ ,  $\omega \in \mathbb{R}_+$ , and  $\hat{q} \in \mathbb{C}^n$  be given. Assume  $\sigma$  and  $\lambda$  in (10) are well defined. The MHB equation (5) with  $\hat{x} = \xi_\ell$  is satisfied if and only if

$$\eta_i \kappa(|\hat{q}_i|) = \phi_\ell(\omega, \omega_o), \quad (12)$$

$$\theta_\ell(\omega, \omega_o) = 0, \quad (13)$$

$$\hat{q}_i = r(j\omega) h_i \hat{y}_{\ell i}, \quad (14)$$

hold for all  $i \in \mathbb{I}_n$ , where  $\kappa$  is defined by (4),  $\hat{y}_{\ell i}$  is the  $i^{\text{th}}$  entry of  $\hat{y}_\ell := C\xi_\ell$ , and

$$\begin{aligned} \eta_i &:= g_i h_i / k_i, \\ \phi_\ell(\omega, \omega_o) &:= -(2\omega/\omega_\ell)(\sigma\varpi_\ell + \lambda\zeta_\ell), \\ \theta_\ell(\omega, \omega_o) &:= (2\omega/\omega_\ell)(\sigma\zeta_\ell - \lambda\varpi_\ell). \end{aligned}$$

For a given controller, the frequency of oscillation with a mode shape  $\xi_\ell$ , if any, can be estimated as a solution to (13). The estimated frequency, denoted by  $w_\ell$ , for specific choices of  $r(s)$  is summarized in Table I. We see that  $w_\ell$  is exactly equal to  $\omega_\ell$  when  $r(s)$  is D or I. For LPF (respectively HPF),  $w_\ell$  is greater (smaller) than  $\omega_\ell$ , but is close to  $\omega_\ell$  when the damping  $\rho$  is small and/or  $\omega_o$  is small (large). For BPF condition (13) reduces to  $\zeta_\ell = \varpi_o \varpi_\ell$ , the right hand side of which can be viewed as a function of  $\omega$  that vanishes at  $\omega = \omega_o$  and  $\omega_\ell$  and is convex on  $\omega > 0$ . Hence, (13) has two distinct real positive solutions, denoted by  $\omega = w_\ell^\pm$ , such that  $\omega_o$  and  $\omega_\ell$  are contained in the interval  $w_\ell^- < \omega < w_\ell^+$ . When the damping  $\zeta_\ell$  is small, one of  $w_\ell^\pm$  is close to  $\omega_o$ , and the other is close to  $\omega_\ell$ . Explicit formulas for  $w_\ell^\pm$  are easily derived, but are not shown for brevity.

Condition (12) suggests a bound on the feedback gain  $\eta_i$  so that the MHB equation admits a solution and predicts an oscillation at  $(\omega, \xi_\ell)$ . In particular, the describing function

TABLE I  
ESTIMATED FREQUENCY AND GAIN BOUND

$r(s)$	$w_\ell$	$\phi_\ell(w_\ell, \omega_o)$
D	$\omega_\ell$	$\rho$
I	$\omega_\ell$	$-\rho\omega_\ell^2$
LPF	$\omega_\ell\sqrt{1+\rho\omega_o}$	$-\rho\omega_o(1+(w_\ell/\omega_o)^2)$
HPF	$\omega_\ell/\sqrt{1+\rho\omega_\ell^2/\omega_o}$	$\rho\omega_o(1+(w_\ell/\omega_o)^2)$

$\kappa(|\hat{q}_i|)$  takes a value in the interval  $0 < \kappa(|\hat{q}_i|) < 1$ , and hence (12) implies  $0 < \phi_\ell(\omega, \omega_o)/\eta_i < 1$ . Thus, if the controller assigns  $(\omega, \xi_\ell)$ , then it is necessary (and sufficient) that  $\eta_i$  have the same sign as  $\phi_\ell(\omega, \omega_o)$  and its absolute value be greater than  $|\phi_\ell(\omega, \omega_o)|$ .

*Theorem 1:* Consider the mechanical system (1) and controller (2), where  $\psi \in \Psi$  and  $r(s) \in \mathcal{R}$ . Let  $\ell \in \mathbb{I}_n$  and  $\omega \in \mathbb{R}_+$  be given. The controller assigns  $(\omega, \xi_\ell)$  if and only if there exists  $\eta \in \mathbb{R}^n$  such that

$$g_i = \frac{k_i \eta_i}{h_i}, \quad h_i = \frac{\alpha_i}{|r(j\omega)\hat{y}_{\ell i}|}, \quad (15)$$

where  $\alpha_i$  is the real positive number satisfying  $\kappa(\alpha_i) = \phi_\ell(\omega, \omega_o)/\eta_i$ , and, for all  $i \in \mathbb{I}_n$ ,

- (i)  $\omega = w_\ell$  and  $\eta_i > \phi_\ell(\omega, \omega_o)$  for D and HPF;
- (ii)  $\omega = w_\ell$  and  $\eta_i < \phi_\ell(\omega, \omega_o)$  for I and LPF;
- (iii)  $\omega = w_\ell^-$  and  $\eta_i > \phi_\ell(\omega, \omega_o)$ , or  
 $\omega = w_\ell^+$  and  $\eta_i < \phi_\ell(\omega, \omega_o)$ , for BPF.

For each case, the sign of the gain bound  $\phi_\ell(\omega, \omega_o)$  is such that  $\eta_i = 0$  is excluded from the feasible set of  $\eta_i$ .

*Proof:* From Lemma 1, the controller assigns  $(\omega, \xi_\ell)$  if and only if (12) and (13) are satisfied with  $\hat{q}_i$  defined by (14). Following the preceding development, one can verify that (12) and (13) are satisfied if and only if the given conditions are satisfied. ■

Theorem 1 suggests that  $(\eta, \omega_o)$  are the essential controller parameters where  $\eta := [\eta_1, \dots, \eta_n]^\top$ . The frequency  $\omega$  of the closed-loop oscillation is determined by  $\omega_o$  through (13). For a given  $\eta$ , the parameters  $(g, h)$  that assign  $(\omega, \xi_\ell)$  are obtained by solving (12) for  $|\hat{q}_i|$  ( $= \alpha_i$ ), and (14) for  $h_i$ , and finally letting  $g_i := k_i \eta_i / h_i$ .

### B. Stability Analysis and Existence of Oscillations

The quasi-linear system (6) depends on the product of  $g_i$ ,  $h_i$  and  $\kappa(\alpha_i)$ , but not on the individual values, and  $\alpha_i$  is determined by  $(\eta, \omega_o)$  for a chosen mode  $\ell$  through Theorem 1. Hence, the stability property of (6) is determined by  $\ell$  and  $(\eta, \omega_o)$ . Due to the special structure of (1) satisfying Assumption 1, pole locations of the associated linear system (6) can be examined through the Routh stability criterion. The following result shows when (6) satisfies the marginal stability condition.

*Theorem 2:* Consider the mechanical system (1) and controller (2), where  $\psi \in \Psi$  and  $r(s) \in \mathcal{R}$ . Let  $\ell \in \mathbb{I}_n$  and  $\omega \in \mathbb{R}_+$  be given and suppose the controller assigns  $(\omega, \xi_\ell)$ . Then the following stability property holds for each controller:

- (i) All the poles of (6) are on the imaginary axis for D;

- (ii) All the poles except  $s = \pm j\omega$  of (6) are in the open left half plane if and only if  $\ell = 1$ , for I, LPF, and HPF;
- (iii) All the poles except  $s = \pm j\omega$  of (6) are in the open left half plane if and only if

$$\phi_i(w_i^+, \omega_o) < \phi_\ell(\omega, \omega_o) < \phi_i(w_i^-, \omega_o) \quad (16)$$

for all  $i \in \mathbb{I}_n / \{\ell\}$  where  $w = w_i^\pm$  are the positive solutions of  $\theta_i(w, \omega_o) = 0$ .

In the case of D, all the poles of (6) are on the imaginary axis, and our method does not suggest whether the assigned oscillation  $(\omega, \xi_\ell)$  is stable. For I, LPF and HPF, Theorem 2 shows that only the first natural mode can be stably assigned. For BPF, the condition in (16) means that the selected mode  $\ell \in \mathbb{I}_n$  can be stably assigned if and only if it gives the the smallest bound  $|\phi_\ell(\omega, \omega_o)|$  on the magnitude of the feedback gain  $|\eta|$ , where we note that  $\phi_i(w_i^+, \omega_o) < 0$  and  $\phi_i(w_i^-, \omega_o) > 0$  always hold.

Theorem 1 characterizes the set of controllers that assign a given oscillation profile  $(\omega, \xi_\ell)$  as a solution to the MHB equation. Theorem 2 provides additional conditions under which the assignment is done in a stable manner. In this way, one can design a controller so that the closed-loop system is likely to have an oscillation at  $(\omega, \xi_\ell)$ . However, existence of such oscillation is not guaranteed due to the approximations involved with the MHB method. The following result shows that the existence of Y-oscillation [10], [12], [13] is indeed guaranteed when the same value is used for the feedback gains  $\eta_i$  in every input-output channel, under a mild assumption on the nature of the equilibrium. Note that Y-oscillation is less restrictive than a periodic trajectory; that is, a periodic trajectory is a Y-oscillation but the converse is not necessarily true.

*Theorem 3:* Let a mechanical system (1) and controller (2) be given, where  $\psi \in \Psi$  and  $r(s) \in \mathcal{R}$ . Suppose the origin of the closed-loop system (3) is a hyperbolic equilibrium, and there exist  $\eta_o \in \mathbb{R}$  such that  $g_i h_i = k_i \eta_o$  holds for all  $i \in \mathbb{I}_n$ , and  $\eta_i := \eta_o$  satisfies statements (i)–(iii) in Theorem 1 for some  $\ell \in \mathbb{I}_n$ . Then, for almost all initial conditions, the trajectory  $z(t)$  of (3) oscillates in the steady state.

This theorem is proved by showing that (a) the equilibrium at the origin is unique and unstable, and (b) every trajectory is bounded. The requirement that  $\eta_i := g_i h_i / k_i$  be independent of  $i$  is introduced for the sake of tractability, and hence may not be necessary to conclude existence of oscillation. However, such choice may be reasonable because it achieves signal conditioning by making the magnitude of the input signal  $q_i$  to the nonlinearity  $\psi$  uniform over every agents.

In summary, a controller can be designed to assign  $(\omega, \xi_\ell)$  as a (stable) oscillation profile by first selecting a natural mode  $\ell$  and parameters  $(\omega_o, \eta_o)$  so that  $0 < \phi_\ell(\omega, \omega_o)/\eta_o < 1$  where  $\omega := w_\ell^+$  or  $w_\ell^-$  for BPF, and  $\omega := w_\ell$  for the others, and then setting the feedback gains  $g_i$  and  $h_i$  by (15). This general procedure applies to all the cases of  $r(s) \in \mathcal{R}$ , with the following specific details:

- For D, the natural mode  $(\omega_\ell, \xi_\ell)$  is assigned exactly when  $\eta_o > \rho$ , but the resulting oscillation may not be stable.

- For I, the natural mode  $(\omega_\ell, \xi_\ell)$  is assigned exactly when  $\eta_o < -\rho\omega_\ell^2$ , and the resulting oscillation is expected to be stable when the first mode ( $\ell = 1$ ) is selected.
- For LPF and HPF, the natural mode  $(\omega_\ell, \xi_\ell)$  is assigned approximately ( $\omega \cong \omega_\ell$ ) when

$$\begin{aligned} \text{LPF: } \eta_o &> -\rho\omega_o(1 + (w_\ell/\omega_o)^2), \\ \text{HPF: } \eta_o &< \rho\omega_o(1 + (w_\ell/\omega_o)^2), \end{aligned} \quad (17)$$

and the resulting oscillation is expected to be stable when the first mode ( $\ell = 1$ ) is selected.

- For BPF, the natural mode  $(\omega_\ell, \xi_\ell)$  is assigned approximately ( $\omega \cong \omega_\ell$ ) when either

$$\begin{aligned} \eta_o &< \phi_\ell(w_\ell^+, \omega_o), \quad \omega_o \leq \omega_\ell, \quad \text{or} \\ \eta_o &> \phi_\ell(w_\ell^-, \omega_o), \quad \omega_o \geq \omega_\ell. \end{aligned} \quad (18)$$

The oscillation profile  $(\omega_o, \xi_\ell)$  is assigned approximately ( $\omega \cong \omega_o$ ) when either

$$\begin{aligned} \eta_o &< \phi_\ell(w_\ell^+, \omega_o), \quad \omega_o \geq \omega_\ell, \quad \text{or} \\ \eta_o &> \phi_\ell(w_\ell^-, \omega_o), \quad \omega_o \leq \omega_\ell. \end{aligned} \quad (19)$$

In each case, the resulting oscillation is expected to be stable when (16) holds with  $\omega = w_\ell^+$  for negative  $\eta_o$  or  $w_\ell^-$  for positive  $\eta_o$ .

In the above, the approximate assignment ( $\omega \cong \omega_\ell$  or  $\omega \cong \omega_o$ ) has smaller error if the damping  $\rho$  is smaller.

The control designs for D and I are simple; just choose a positive (for D) or negative (for I) feedback gain  $\eta_o$  with its magnitude large enough. For LPF, HPF, and BPF, the design parameters  $(\omega_o, \eta_o)$  can be selected using the mode-partition diagram as shown in Fig. 1. In this diagram, the parameter plane  $(\omega_o, \eta_o)$  can be partitioned into multiple regions  $\mathbb{O}_{\ell\ell}$ ,  $\mathbb{O}_{o\ell}$ , and  $\mathbb{O}_{oo}$ , where oscillation profiles  $(\omega_\ell, \xi_\ell)$  and  $(\omega_o, \xi_\ell)$  are assigned approximately but stably in  $\mathbb{O}_{\ell\ell}$  and  $\mathbb{O}_{o\ell}$ , respectively, and no oscillation is assigned stably in  $\mathbb{O}_{oo}$  because the origin is stable. Specifically, for LPF and HPF,  $\mathbb{O}_{11}$  is defined by (17) with  $\ell = 1$  and the complement of this region is  $\mathbb{O}_{oo}$ . For BPF,  $\mathbb{O}_{\ell\ell}$  is defined by (18) and (16),  $\mathbb{O}_{o\ell}$  is defined by (19) and (16), and the remaining region is  $\mathbb{O}_{oo}$ .

#### IV. DESIGN EXAMPLES

We give design examples to illustrate the proposed method for natural entrainment and to complement the approximate nature of the theoretical conditions using numerical simulations. We consider a 3-link mechanical arm with flexible joints on a horizontal plane, where the first link is connected to the inertial frame through a rotational joint. The  $i^{\text{th}}$  link has uniformly distributed mass  $m_i$ , length  $\ell_i$ , and the moment of inertia  $J_i := m_i\ell_i^2/12$ . Mounted at the  $i^{\text{th}}$  joint are a spring of stiffness  $k_i$ , a dashpot of damping coefficient  $\rho k_i$ , and an actuator that generates torque input  $u_i$ . The equation of motion linearized at the origin is given by (1) where  $x_i$  is the angular displacement of the  $i^{\text{th}}$  link,  $y_i$  is the relative angle

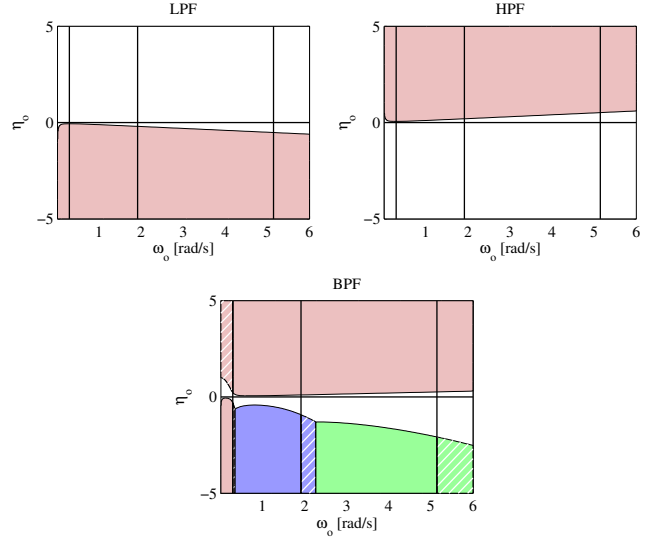


Fig. 1. Mode-partition Diagrams for 3-link Mechanical Arm: ( $\omega_1 = 0.292$ ,  $\omega_2 = 1.92$ ,  $\omega_3 = 5.15$ ); solid pink:  $\mathbb{O}_{11}$ , shaded pink:  $\mathbb{O}_{o1}$ , solid blue:  $\mathbb{O}_{22}$ , shaded blue:  $\mathbb{O}_{o2}$ , solid green:  $\mathbb{O}_{33}$ , shaded green:  $\mathbb{O}_{o3}$ , where in the region labeled as  $\mathbb{O}_{pq}$ , the MHB analysis predicts that the closed-loop oscillation has a frequency close to the  $p^{\text{th}}$  natural frequency  $\omega_p$  (or  $\omega_o$  if  $p = 0$ ) with the  $q^{\text{th}}$  natural mode shape  $\xi_q$ .

between the  $i^{\text{th}}$  and  $(i+1)^{\text{th}}$  link, and

$$\begin{aligned} J &:= \text{diag}(J_1, J_2, J_3) + L^T M L, \\ M &:= \text{diag}(m_1, m_2, m_3), \quad \mathcal{K} := \text{diag}(k_1, k_2, k_3), \\ C &:= B^T, \quad K := B\mathcal{K}C, \quad D := \rho K, \\ B &:= \begin{bmatrix} 1 & -1 & 0 \\ 0 & 1 & -1 \\ 0 & 0 & 1 \end{bmatrix}, \quad L := \begin{bmatrix} \ell_1/2 & 0 & 0 \\ \ell_1 & \ell_2/2 & 0 \\ \ell_1 & \ell_2 & \ell_3/2 \end{bmatrix}. \end{aligned}$$

We use the following parameters for the mechanical system:

$$\ell_i = 1, \quad m_i = 1, \quad k_i = 1, \quad \rho = 0.1,$$

for  $i \in \mathbb{I}_3$ . The natural modes are given by

$\ell$	1 <sup>st</sup> mode	2 <sup>nd</sup> mode	3 <sup>rd</sup> mode
$\omega_\ell$	0.292	1.92	5.15
$\xi_\ell$	0.402	0.476	0.291
	0.615	-0.242	-0.654
	0.678	-0.846	0.698

where the mode shape  $\xi_\ell$  is normalized so that  $\|\xi_\ell\| = 1$ .

For the control design, we use the nonlinearity  $\psi(x) = \tanh(x)$ . The feedback gain  $\eta_o$  is selected to satisfy the bound  $\eta_o > \rho$  and  $\eta_o < -\rho\omega_\ell^2$  for D and I, respectively, and  $(\omega_o, \eta_o)$  is selected from the mode-partition diagrams in Fig. 1 for LPF, HPF, and BPF. The feedback gains  $(g, h)$  are then calculated from (15). For each controller thus designed, numerical simulations are performed for the system and accuracy of the MHB method is evaluated with the following error criteria:

$$\begin{aligned} e_{\text{PRD}} &:= |\omega_{\text{MHB}} - \omega_{\text{SIM}}|/\omega_{\text{SIM}}, \quad e_\omega := |\omega_{\text{SIM}} - \omega_{\text{NAT}}|/\omega_{\text{NAT}}, \\ e_x &:= \|\bar{x}_{\text{SIM}} - x_{\text{NAT}}\|/\|x_{\text{NAT}}\|, \quad \bar{x}_{\text{SIM}} := x_{\text{SIM}}/\|x_{\text{SIM}}\|, \end{aligned}$$

where  $\omega_{\text{MHB}}$ ,  $\omega_{\text{SIM}}$ , and  $\omega_{\text{NAT}}$  are the oscillation frequency predicted by the MHB method, the actual frequency achieved

in the simulations, and the natural frequency targeted in the design, respectively, and  $x_{\text{SIM}}$  and  $x_{\text{NAT}}$  are defined similarly for the mode shape. Below, the results for the error values are reported in the unit of %.

For D, we designed three controllers, one for each mode, and the result is summarized below.

	$\xi_\ell$	$\eta$	$\omega_{\text{MHB}}$	$\omega_{\text{SIM}}$	$e_{\text{PRD}}$	$e_\omega$	$e_x$
(a)	$\xi_1$	3	0.292	1.90	84	551	175
(b)	$\xi_2$	3	1.92	1.91	0.4	0.4	0.0
(c)	$\xi_3$	3	5.15	4.99	3.2	3.1	0.0

While our theoretical result is inconclusive for D regarding which mode can be stably entrained, numerical simulations revealed that the second and third modes could be entrained but not the first mode.

For I, we obtained the following result.

$\xi_\ell$	$\eta$	$\omega_{\text{MHB}}$	$\omega_{\text{SIM}}$	$e_{\text{PRD}}$	$e_\omega$	$e_x$
$\xi_1$	-3	0.292	0.292	0.0	0.0	0.0

As predicted in Theorem 2, only the first mode can be stably entrained. The errors  $e_{\text{PRD}}$ ,  $e_\omega$  and  $e_x$  are all small, indicating accuracy of the MHB method.

For LPF, we designed five controllers with different values of  $\omega_o$  and obtained the following results.

	$\xi_\ell$	$\omega_o$	$\eta$	$\omega_{\text{MHB}}$	$\omega_{\text{SIM}}$	$e_{\text{PRD}}$	$e_\omega$	$e_x$
(a)	$\xi_1$	0.2	-3	0.295	0.295	0.0	1.0	0.0
(b)	$\xi_1$	1.0	-3	0.307	0.307	0.0	4.8	0.0
(c)	$\xi_1$	4.0	-3	0.346	0.344	0.6	17	0.0
(d)	$\xi_1$	6.0	-3	0.370	0.367	0.9	25	0.0

In accordance with the mode-partition diagram in Fig. 1, the first mode was entrained regardless of the value of  $\omega_o$ . The errors  $e_{\text{PRD}}$  and  $e_x$  were found to be small, but  $\omega_{\text{SIM}}$  deviates from  $\omega_1$  as  $\omega_o$  gets larger as predicted in Table I.

For HPF, we designed four controllers with the results given below.

	$\xi_\ell$	$\omega_o$	$\eta$	$\omega_{\text{MHB}}$	$\omega_{\text{SIM}}$	$e_{\text{PRD}}$	$e_\omega$	$e_x$
(a)	$\xi_1$	0.2	3	0.286	0.286	0.0	2.1	0.0
(b)	$\xi_1$	1.0	3	0.291	0.291	0.0	0.4	0.0
(c)	$\xi_1$	4.0	3	0.292	0.292	0.0	0.1	0.0
(d)	$\xi_1$	6.0	3	0.292	0.292	0.0	0.1	0.0

The results are similar to the LPF case; only the first mode was entrained, and the errors are small, with  $e_\omega$  decreasing with  $\omega_o$  as predicted from Table I.

For BPF, we obtained the results below.

	$\xi_\ell$	$\omega_o$	$\eta$	$\omega_{\text{MHB}}$	$\omega_{\text{SIM}}$	$e_{\text{PRD}}$	$e_\omega$	$e_x$
(a)	$\xi_1$	1.0	3	0.290	0.290	0.0	0.9	0.0
(b)	$\xi_1$	4.0	3	0.292	0.292	0.0	0.2	0.0
(c)	$\xi_1$	6.0	3	0.292	0.292	0.0	0.2	0.0
(d)	$\xi_1$	0.2	-3	0.303	0.303	0.0	3.5	0.0
(e)	$\xi_2$	1.0	-3	2.15	2.14	0.1	12	0.0
(f)	$\xi_3$	4.0	-3	7.49	7.48	0.1	45	0.0

As predicted from the mode-partition diagram in Fig. 1, only the first mode is entrained by a positive feedback, regardless of the value of  $\omega_o$ . In contrast, every specified mode can

be entrained by a negative feedback with an appropriately selected  $\omega_o$ . For both positive and negative feedbacks,  $e_{\text{PRD}}$  and  $e_x$  are small, but  $e_\omega$  is relatively large for (d)–(f). The deviation from  $\omega_\ell$  is predicted by the MHB method; the estimated frequency  $w_\ell^\pm$  is close to  $\omega_\ell$  or  $\omega_o$  when the damping ratio  $\zeta_\ell := \rho\omega_\ell/2$  is small and the distance between  $\omega_o$  and  $\omega_\ell$  is large.

## V. CONCLUSION

We have investigated the possibility of achieving entrainment to a natural mode of oscillation for multi-DOF collocated mechanical systems using multi-agent controllers, each of which is placed in a feedback loop between a sensor/actuator pair. Necessary and sufficient conditions are obtained for the controller so that the harmonic balancing predicts a stable oscillation near a prescribed natural mode. The analytical results indicated, and numerical simulations confirmed, that multiple agents with simple dynamics are able to achieve natural entrainment, and that the band-pass filtering property of agent dynamics is important for entrainment to an arbitrarily specified mode. The mode-partition diagram for the BPF agent case suggests, however, that the BPF agent may not fully capture the dynamical properties of the CPG “agent” used in [8] that are important for *robust* entrainment against perturbations in mechanical parameters.

## REFERENCES

- [1] F. Delcomyn. Neural basis of rhythmic behavior in animals. *Science*, 210:492–498, 1980.
- [2] G. N. Orlovsky, T.G. Deliagina, and S. Grillner. *Neuronal Control of Locomotion: From Mollusc to Man*. Oxford University Press, 1999.
- [3] T. Iwasaki and M. Zheng. Sensory feedback mechanism underlying entrainment of central pattern generator to mechanical resonance. *Biological Cybernetics*, 94(4):245–261, 2006.
- [4] B. W. Verdaasdonk, H. F. Koopman, and F. C. van der Helm. Energy efficient and robust rhythmic limb movement by central pattern generators. *Neural Network*, 19(4):388–400, 2006.
- [5] B. W. Verdaasdonk, H. F. Koopman, and F. C. van der Helm. Resonance tuning in a neuro-musculo-skeletal model of the forearm. *Biological Cybernetics*, 96(2):165–180, 2007.
- [6] C. A. Williams and S. P. DeWeerth. A comparison of resonance tuning with positive versus negative sensory feedback. *Biological Cybernetics*, 96(6):603–614, 2007.
- [7] Y. Futakata and T. Iwasaki. Formal analysis of resonance entrainment by central pattern generator. *J. Math. Biol.*, 57(2):183–207, 2008.
- [8] Y. Futakata and T. Iwasaki. Entrainment of central pattern generators to natural oscillations of collocated mechanical systems. In *IEEE Conf. Decision and Contr.*, pages 5220–5225, Cancun, Mexico, December 2008.
- [9] Y. Futakata. *Natural mode entrainment by CPG-based decentralized feedback controllers*. Ph.D. dissertation, Mechanical and Aerospace Engineering, University of Virginia, August 2009.
- [10] T. Iwasaki. Multivariable harmonic balance for central pattern generators. *Automatica*, 44(12):4061–4069, 2008.
- [11] Z. Chen, M. Zheng, W. O. Friesen, and T. Iwasaki. Multivariable harmonic balance analysis of neuronal oscillator for leech swimming. *J. Computational Neuroscience*, 25(3):583–606, 2008. (DOI: 10.1007/s10827-008-0105-7).
- [12] V. A. Yakubovich. Frequency-domain criteria for oscillation in nonlinear systems with one stationary nonlinear component. *Siberian Mathematical Journal*, 14(5):768–788, 1973.
- [13] A. Pogromsky, T. Glad, and H. Nijmeijer. On diffusion driven oscillations in coupled dynamical systems. *Int. J. Bifurcation and Chaos*, 9(4):629–644, 1999.